

Estimation of Overlapping Modes using Sparse Modeling of  
Signal Innovation

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## Introduction

- Focus on multi-component signals (MCS).

$$x(n) = \sum_{k=0}^{K-1} x_k(n) \quad , \quad \text{with } x_k(n) = a_k(n) e^{j\phi_k(n)}, \quad (1)$$

### Investigated approaches

- Mixture of  $K$  superimposed components.
- $a_k(n)$  and  $\phi_k(n)$  the time-varying amplitude and frequency of component  $k$ .

## Problem statement

### Objective

- Instantaneous frequency (IF) estimation.
- Overlapping components.
- Presence of noise.

### Context of this work

- Time-frequency representation (TFR).
- Well design representation.
- Allows to observe the IF trajectory of each mode as a ridge.

### Challenge

- Large variety of real signals (amplitude, modulation rate, . . . ).
- Presence of external noise.
- Vicinity of ridges overlap.

# Plan

- 1 Observation model
- 2 Estimation strategy
- 3 Results
- 4 Conclusion

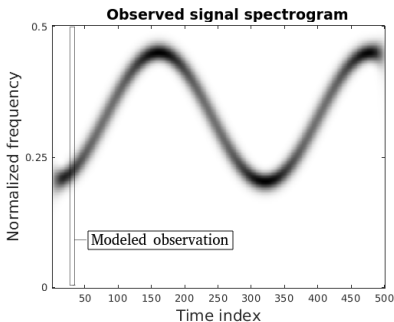
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## Observation model

### Challenge

- Spectrogram of the observation: square modulus of the STFT.
- Model vertical signal spectrogram cut.
- 1D signal observed at a fixed time instant.

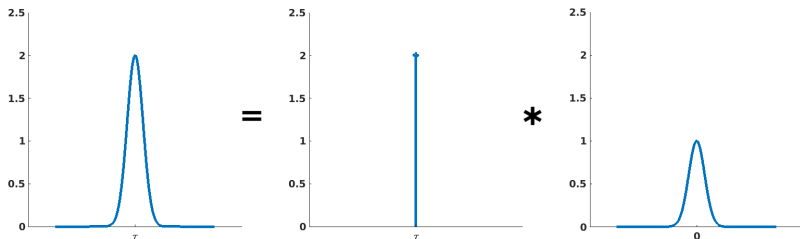


- $g(m) = e^{-\left(\frac{2\pi mL}{M}\right)^2}$ .
- $m$  : frequency in  $[0, M - 1]$ .
- $L$  : time spread of the analysis window

- Known Gaussian analysis window.
- Interest in its shift parameter.
- Indicates the position of each ridges.

## Observation model

- Observation can be modeled as a discrete convolution between a stream of Dirac pulses and  $g$ .
- Known filter  $g \rightarrow$  square of the Fourier transform of the analysis window.
- Pulses position (resp. weight) inform on the ridges location (resp. amplitude).



## Observation model

- For a given stream of Dirac pulses associated with time instant  $n$

$$f_n(m) = \sum_{k=0}^{K-1} a_k(n) \delta(m - \phi'_k(n)), \quad (2)$$

with  $\phi'_k(n) = \frac{d\phi_k}{dn}(n)$  the IF, we observe

$$s_n(m) = \sum_{k=0}^{K-1} a_k(n) g(m - \phi'_k(n)). \quad (3)$$

- $f_n$  is the most common example of signal with a finite rate of innovation (FRI).



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## Classical reconstruction

### Sparse reconstruction

- Reconstructing FRI signal from a filtered and sampled version of them is a known problem<sup>1</sup>.
- Let us write  $s_n$  as function of the inverse discrete time Fourier series (DTFS) of  $g$ .

$$\begin{aligned}
 s_n(m) &= \sum_{k=0}^{K-1} a_k(n) \sum_{\lambda=-\infty}^{\infty} \hat{g}(\lambda) e^{\frac{j2\pi\lambda(m-\phi'_k(n))}{M}} \\
 &= \sum_{\lambda=-\infty}^{\infty} \hat{g}(\lambda) \underbrace{\sum_{k=0}^{K-1} a_k(n) e^{\frac{-j2\pi\lambda\phi'_k(n)}{M}}}_{\hat{f}_n(\lambda)} e^{\frac{j2\pi\lambda m}{M}}, \tag{4}
 \end{aligned}$$

with  $\hat{g}$  (resp.  $\hat{f}_n$ ) the Fourier transform of  $g$  (resp.  $f_n$ ).

- Involve an infinite sum.

<sup>1</sup>M. Vetterli and P. Marziliano and T. Blu. Sampling signals with finite rate of innovation, 2002.

## Classical reconstruction

### Sparse reconstruction

- Bandlimited approximation: keep only  $2M_0 + 1$  Fourier series coefficients<sup>2</sup>.

$$s_n(m) \approx \sum_{\lambda=-M_0}^{M_0} \hat{g}(\lambda) \hat{f}_n(\lambda) e^{\frac{j2\pi\lambda m}{M}}. \quad (5)$$

- We can invert the matrix wise system

$$s_n = \mathbf{V} \mathbf{D}_g \hat{f}_n \Leftrightarrow \hat{f}_n = \mathbf{D}_g^{-1} \mathbf{V}^{-1} s_n, \quad (6)$$

- $\mathbf{V}$  is a Vandermonde matrix (Fourier atoms).
- $\mathbf{D}_g$  is diagonal and gathers the DTFS coefficients  $\hat{g}(\lambda)$ .

<sup>2</sup>A. Bhandari and A. M. Wallace and R. Raskar. Super-resolved time-of-flight sensing via FRI sampling theory, 2016.

## Alternative reconstruction

### Pulses location and weight recovery

- At this point, we should use the Prony method.
- Consists in computing a filter  $h$  that annihilates  $f_n$ .

$$\underbrace{\begin{pmatrix} \hat{f}_n(0) & \cdots & \hat{f}_n(-K+1) \\ \hat{f}_n(1) & \cdots & \hat{f}_n(-K+2) \\ \vdots & \vdots & \vdots \\ \hat{f}_n(K-1) & \cdots & \hat{f}_n(0) \end{pmatrix}}_A \begin{pmatrix} h(1) \\ h(2) \\ \vdots \\ h(K) \end{pmatrix} = - \begin{pmatrix} \hat{f}_n(1) \\ \hat{f}_n(2) \\ \vdots \\ \hat{f}_n(K) \end{pmatrix} \quad (7)$$

- The roots of the Z-transform of  $h$  are the  $e^{\frac{-j2\pi\phi_k'(n)}{M}}$ .
- Prony method is limited in the presence of noise<sup>1</sup>.

<sup>1</sup>M. Vetterli and P. Marziliano and T. Blu. Sampling signals with finite rate of innovation, 2002.

## Alternative reconstruction

### Presence of noise

- Inverting the Vandermonde matrix becomes an ill-posed problem.
- Computation of a filter  $h$  that does not annihilate  $f_n$  anymore.
- Existing alternatives.

### Alternative

- Replacing the Prony method.
- Total Least-Squares alternative<sup>3</sup>.
- Do not search anymore for a filter that annihilates  $f_n$ .
- Estimate a filter  $h$  minimizing  $\|\mathbf{A}h\|^2$  under the constraint  $\|h\|_2 = 1$ .

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<sup>3</sup>T. Blu and P. L. Dragotti and M. Vetterli and P. Marziliano and L. Coulot. Sparse sampling of signal innovations, 2008.

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## Results

### Experiments

- Comparison to a pseudo-Bayesian (PB) approach<sup>4</sup>.
- Using several variational objectives (divergences, hyperparameters).
- Comparison of FRI (classical) and FRI\_TLS (proposed).

$$\text{RMSE} = \sum_{k=0}^K \left[ \sum_{n=0}^{N-1} \frac{(\bar{m}_{n,k} - \hat{m}_{n,k})^2}{M^2} \right], \quad (8)$$

- $\bar{m}_{n,k}$  (resp.  $\hat{m}_{n,k}$ ) the actual (resp. estimated) ridge position of the  $k$ -th component in the  $n$ -th time instant.
- White Gaussian noise with various Signal-to-noise ratio (SNR).

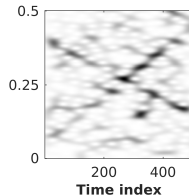
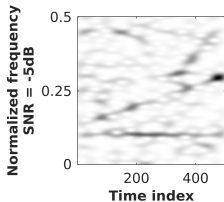
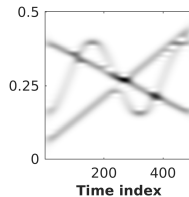
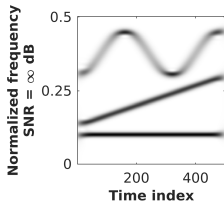
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<sup>4</sup>Q. Legros, D. Fourer. A novel pseudo-Bayesian approach for robust multi-ridge detection and mode retrieval, 2020.

## Results

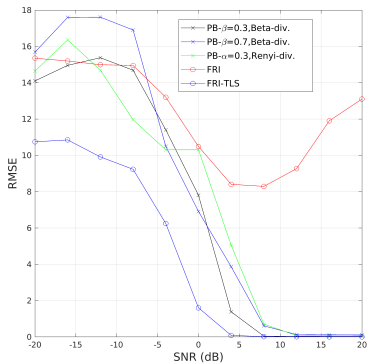
### Numerical experiments

- Reconstruction of two MCSs.
- With and without components overlapping.





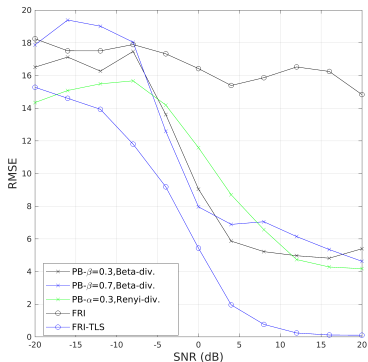
## Results - without overlap



### Observations

- Proposed FRI-TLS provides the best performance for all SNRs.
- FRI-TLS performs almost perfect estimation for  $\text{SNR} > 0\text{dB}$ .
- FRI not efficient for all SNRs.
- Robust to frequency modulation, involving broader ridges.

## Results - with overlap



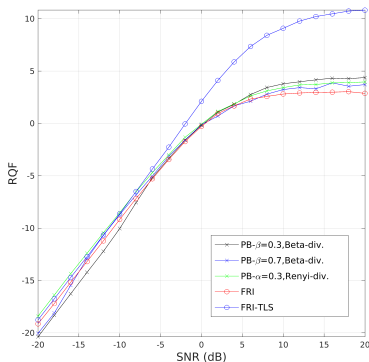
### Observations

- Similar performance.
- FRI-TLS outperforms other approaches except at low SNR.
- Ability of estimating overlapping components (high SNR).

## Results - with overlap

### Experiments

- Reconstruction performance of an overlapping MCS.
- Reconstruction quality factor:  $RQF = 10 \log_{10} \left( \frac{\|x\|^2}{\|x - \hat{x}\|^2} \right)$ .
- $x$  (resp.  $\hat{x}$ ) stands for the reference (resp. estimated) signal.
- Reconstruction using band-pass thresholding around the IF.



### Observations

- Better reconstruction performance for  $SNR > -5dB$  using FRI-TLS.
- Ability to deal with crossed ridges in noisy TFR.
- Robust methods perform better at low SNR due to strong noise in the IF vicinity.

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## Conclusions and perspectives

### Conclusions

- A novel approach for estimating the modes of a MCS in the presence of noise.
- Well suited to scenarios involving overlapping components.
- Improved performance compared to existing methods with and without overlap.

### Future works

- Disentangle estimated components.
- Spatial regularization for cases involving the presence of perturbation around close ridges.

# Thanks for your attention !

Codes available on my GitHub

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